# FAIRCHILD

SEMICONDUCTOR®

# FSAM10SM60A

# SPM<sup>™</sup> (Smart Power Module)

# **General Description**

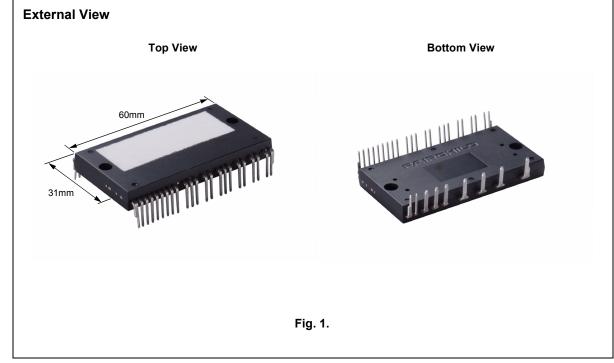
FSAM10SM60A is an advanced smart power module (SPM) that Fairchild has newly developed and designed to provide very compact and high performance ac motor drives mainly targeting medium speed low-power inverterdriven application like air conditioners. It combines optimized circuit protection and drive matched to low-loss IGBTs. Highly effective short-circuit current detection/ protection is realized through the use of advanced current sensing IGBT chips that allow continuous monitoring of the IGBTs current. System reliability is further enhanced by the built-in over-temperature monitoring and integrated undervoltage lock-out protection. The high speed built-in HVIC provides opto-coupler-less IGBT gate driving capability that further reduce the overall size of the inverter system design. In addition the incorporated HVIC facilitates the use of single-supply drive topology enabling the FSAM10SM60A to be driven by only one drive supply voltage without negative bias. Inverter current sensing application can be achieved due to the divided nagative dc terminals.

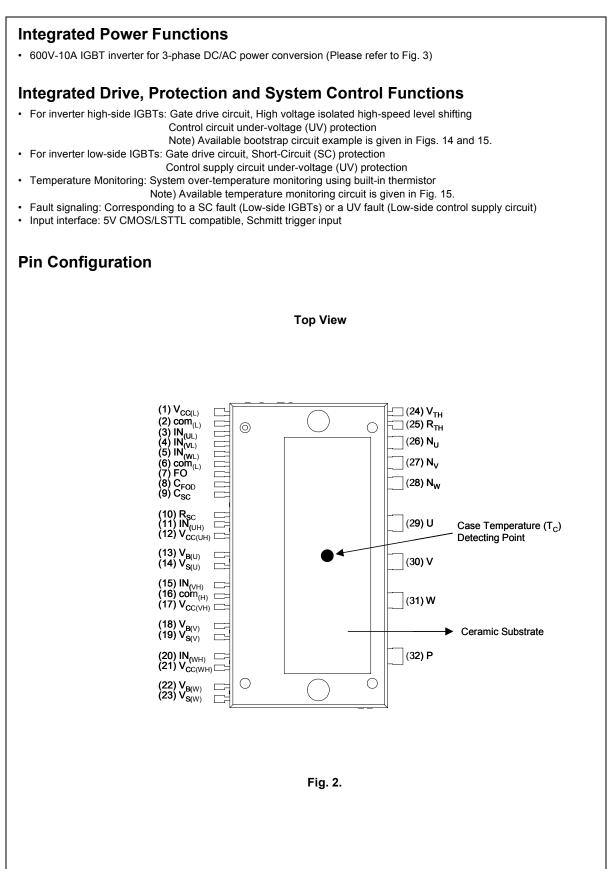
## **Features**

- UL Certified No. E209204
- 600V-10A 3-phase IGBT inverter bridge including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- · Single-grounded power supply due to built-in HVIC
- Typical switching frequency of 5kHz
- · Built-in thermistor for over-temperature monitoring
- Inverter power rating of 0.5kW / 100~253 Vac
- Isolation rating of 2500Vrms/min.
- Very low leakage current due to using ceramic substrate
- Adjustable current protection level by varying series resistor value with sense-IGBTs

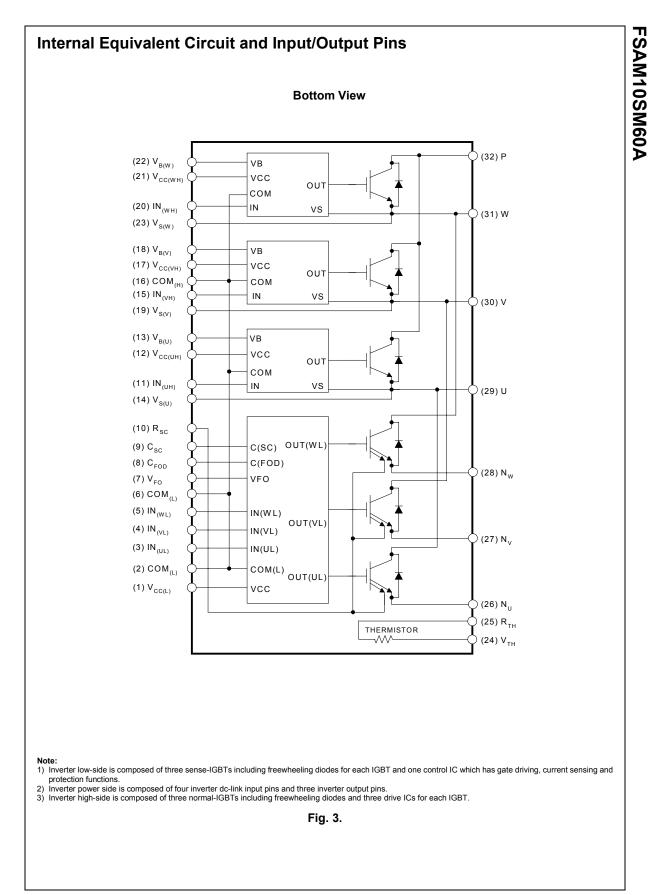
# Applications

- AC 100V ~ 253V 3-phase inverter drive for small power (0.5kW) ac motor drives
- Home appliances applications requiring medium switching frequency operation like air conditioners drive system
- Application ratings:
  - Power : 0.5 kW / 100~253 Vac
  - Switching frequency : Typical 5kHz (PWM Control)
  - 100% load current : 3.3A (Irms)
- 150% load current : 5.0A (Irms) for 1 minute





in Number	Pin Name	Pin Description	
1	V <sub>CC(L)</sub>	Low-side Common Bias Voltage for IC and IGBTs Driving	
2	COM <sub>(L)</sub>	Low-side Common Supply Ground	
3	IN <sub>(UL)</sub>	Signal Input for Low-side U Phase	
4	IN <sub>(VL)</sub>	Signal Input for Low-side V Phase	
5	IN <sub>(WL)</sub>	Signal Input for Low-side W Phase	
6	COM <sub>(L)</sub>	Low-side Common Supply Ground	
7	V <sub>FO</sub>	Fault Output	
8	C <sub>FOD</sub>	Capacitor for Fault Output Duration Time Selection	
9	C <sub>SC</sub>	Capacitor (Low-pass Filter) for Short-Circuit Current Detection Input	
10	R <sub>SC</sub>	Resistor for Short-Circuit Current Detection	
11	IN <sub>(UH)</sub>	Signal Input for High-side U Phase	
12	V <sub>CC(UH)</sub>	High-side Bias Voltage for U Phase IC	
13	V <sub>B(U)</sub>	High-side Bias Voltage for U Phase IGBT Driving	
14	V <sub>S(U)</sub>	High-side Bias Voltage Ground for U Phase IGBT Driving	
15	IN <sub>(VH)</sub>	Signal Input for High-side V Phase	
16	COM <sub>(H)</sub>	High-side Common Supply Ground	
17	V <sub>CC(VH)</sub>	High-side Bias Voltage for V Phase IC	
18	V <sub>B(V)</sub>	High-side Bias Voltage for V Phase IGBT Driving	
19	V <sub>S(V)</sub>	High-side Bias Voltage Ground for V Phase IGBT Driving	
20	IN <sub>(WH)</sub>	Signal Input for High-side W Phase	
21	V <sub>CC(WH)</sub>	High-side Bias Voltage for W Phase IC	
22	V <sub>B(W)</sub>	High-side Bias Voltage for W Phase IGBT Driving	
23	V <sub>S(W)</sub>	High-side Bias Voltage Ground for W Phase IGBT Driving	
24	V <sub>TH</sub>	Thermistor Bias Voltage	
25	R <sub>TH</sub>	Series Resistor for the Use of Thermistor (Temperature Detection)	
26	NU	Negative DC-Link Input for U Phase	
27	N <sub>V</sub>	Negative DC-Link Input for V Phase	
28	N <sub>W</sub>	Negative DC-Link Input for W Phase	
29	U	Output for U Phase	
30	V	Output for V Phase	
31	W	Output for W Phase	
32	Р	Positive DC–Link Input	



# Absolute Maximum Ratings (T<sub>J</sub> = 25°C, Unless Otherwise Specified) **Inverter Part**

Item	Symbol	Condition	Rating	Unit
Supply Voltage	V <sub>PN</sub>	Applied between P- $N_U$ , $N_V$ , $N_W$	450	V
Supply Voltage (Surge)	V <sub>PN(Surge)</sub>	Applied between P- NU, NV, NW	500	V
Collector-Emitter Voltage	V <sub>CES</sub>		600	V
Each IGBT Collector Current	± I <sub>C</sub>	T <sub>C</sub> = 25°C	10	А
Each IGBT Collector Current	± I <sub>C</sub>	T <sub>C</sub> = 100°C	9	А
Each IGBT Collector Current (Peak)	± I <sub>CP</sub>	T <sub>C</sub> = 25°C, Instantaneous Value (Pulse)	20	A
Collector Dissipation	P <sub>C</sub>	T <sub>C</sub> = 25°C per One Chip	43	W
Operating Junction Temperature	ТJ	(Note 1)	-20 ~ 125	°C

Note: 1. It would be recommended that the average junction temperature should be limited to  $T_J \le 125^{\circ}C$  (@ $T_C \le 100^{\circ}C$ ) in order to guarantee safe operation.

# **Control Part**

Item	Symbol	Condition	Rating	Unit
Control Supply Voltage	V <sub>CC</sub>	Applied between $V_{CC(UH)}$ , $V_{CC(VH)}$ , $V_{CC(WH)}$ - $COM_{(H)}$ , $V_{CC(L)}$ - $COM_{(L)}$	20	V
High-side Control Bias Voltage	V <sub>BS</sub>	Applied between V <sub>B(U)</sub> - V <sub>S(U)</sub> , V <sub>B(V)</sub> - V <sub>S(V)</sub> , V <sub>B(W)</sub> - V <sub>S(W)</sub>	20	V
Input Signal Voltage	V <sub>IN</sub>	$ \begin{array}{l} \mbox{Applied between IN}_{(UH)}, \mbox{IN}_{(VH)}, \mbox{IN}_{(WH)} \mbox{-} \mbox{COM}_{(H)} \\ \mbox{IN}_{(UL)}, \mbox{IN}_{(VL)}, \mbox{IN}_{(WL)} \mbox{-} \mbox{COM}_{(L)} \end{array} $	-0.3 ~ V <sub>CC</sub> +0.3	V
Fault Output Supply Voltage	V <sub>FO</sub>	Applied between V <sub>FO</sub> - COM <sub>(L)</sub>	-0.3 ~ V <sub>CC</sub> +0.3	V
Fault Output Current	I <sub>FO</sub>	Sink Current at V <sub>FO</sub> Pin	5	mA
Current Sensing Input Voltage	V <sub>SC</sub>	Applied between C <sub>SC</sub> - COM <sub>(L)</sub>	-0.3 ~ V <sub>CC</sub> +0.3	V

# **Total System**

Item	Symbol	Condition	Rating	Unit
Self Protection Supply Voltage Limit (Short-Circuit Protection Capability)	V <sub>PN(PROT)</sub>	$V_{CC} = V_{BS} = 13.5 \sim 16.5V$ T <sub>J</sub> = 25°C, Non-repetitive, less than 6µs	400	V
Module Case Operation Temperature	Т <sub>С</sub>	Note Fig.2	-20 ~ 100	°C
Storage Temperature	T <sub>STG</sub>		-20 ~ 125	°C
Isolation Voltage	V <sub>ISO</sub>	60Hz, Sinusoidal, AC 1 minute, Connection Pins to Heat-sink Plate	2500	V <sub>rms</sub>

Absolute Maximum Ratings Thermal Resistance							
Item	Symbol	Condition	Min.	Тур.	Max.	Unit	
Junction to Case Thermal Resistance	R <sub>th(j-c)Q</sub>	Each IGBT under Inverter Operating Condition	-	-	2.9	°C/W	
	R <sub>th(j-c)F</sub>	Each FWDi under Inverter Operating Condition	-	-	3.6	°C/W	
Contact Thermal Resistance	R <sub>th(c-h)</sub>	Ceramic Substrate (per 1 Module) Thermal Grease Applied (Note 3)	-	-	0.06	°C/W	

 $\begin{array}{l} \textbf{Note:}\\ \textbf{2. For the measurement point of case temperature(T_C), please refer to Fig. 2.\\ \textbf{3. The thickness of thermal grease should not be more than 100um.} \end{array}$ 

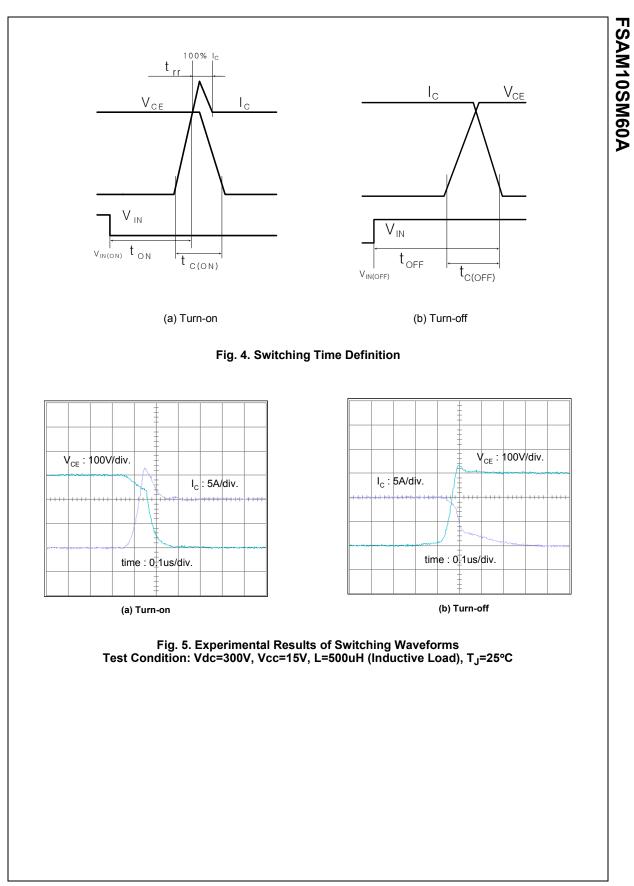
# **Electrical Characteristics** (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

# **Inverter Part**

Item	Symbol	Condition	on	Min.	Тур.	Max.	Unit
Collector - Emitter Saturation Voltage	V <sub>CE(SAT)</sub>	$V_{CC} = V_{BS} = 15V$ $V_{IN} = 0V$	I <sub>C</sub> = 10A, T <sub>J</sub> = 25°C	-	-	2.3	V
FWDi Forward Voltage	V <sub>FM</sub>	V <sub>IN</sub> = 5V	I <sub>C</sub> = 10A, T <sub>J</sub> = 25°C	-	-	2.3	V
Switching Times	t <sub>ON</sub>	V <sub>PN</sub> = 300V, V <sub>CC</sub> = V <sub>BS</sub> = 15	5V	-	0.29	-	us
	t <sub>C(ON)</sub>	I <sub>C</sub> = 10A, T <sub>J</sub> = 25°C		-	0.13	-	us
	t <sub>OFF</sub>	$V_{IN} = 5V \leftrightarrow 0V$ , Inductive Lo (High, Low-side)	ad	-	0.79	-	us
	t <sub>C(OFF)</sub>	(high, Low-side)		-	0.37	-	us
	t <sub>rr</sub>	(Note 4)		-	0.13	-	us
Collector - Emitter Leakage Current	I <sub>CES</sub>	$V_{CE} = V_{CES}, T_J = 25^{\circ}C$		-	-	250	μA

Note:
 t<sub>ON</sub> and t<sub>OFF</sub> include the propagation delay time of the internal drive IC. t<sub>C(ON)</sub> and t<sub>C(OFF)</sub> are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Fig. 4.

# FSAM10SM60A



Rev. E, August 2003

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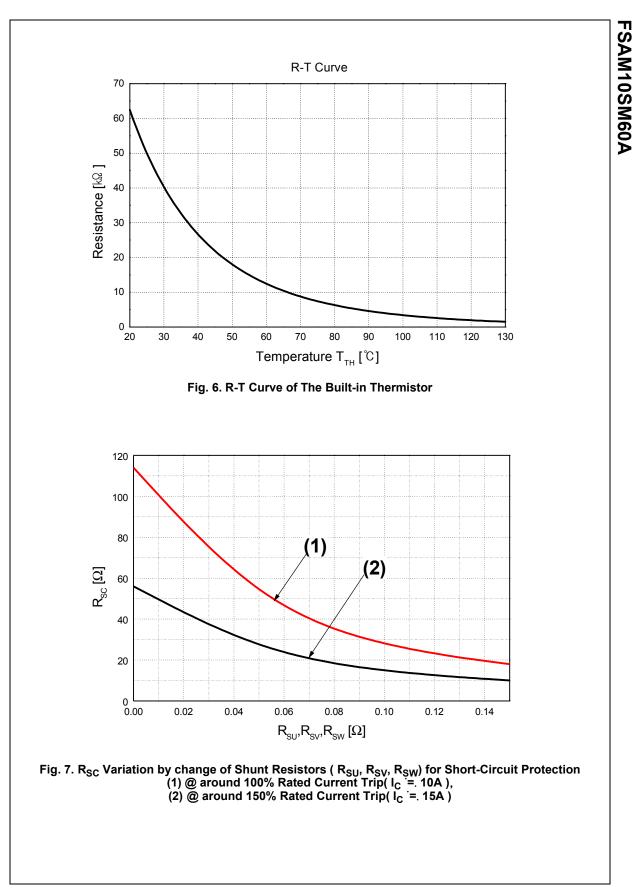
Item	Symbol		Condition	Min.	Тур.	Max.	Unit
Quiescent $V_{CC}$ Supply Current	I <sub>QCCL</sub>	V <sub>CC</sub> = 15V IN <sub>(UL, VL, WL)</sub> = 5V	V <sub>CC(L)</sub> - COM <sub>(L)</sub>	-	-	26	mA
	I <sub>QCCH</sub>	V <sub>CC</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 5V	$V_{CC(UH)}$ , $V_{CC(VH)}$ , $V_{CC(WH)}$ - $COM_{(H)}$	-	-	130	uA
Quiescent V <sub>BS</sub> Supply Cur- rent	I <sub>QBS</sub>	V <sub>BS</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 5V	$V_{B(U)} - V_{S(U)}, V_{B(V)} - V_{S(V)}, V_{B(W)} - V_{S(W)}$	-	-	420	uA
Fault Output Voltage	V <sub>FOH</sub>	$V_{SC} = 0V, V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		4.5	-	-	V
V <sub>FOL</sub>		$V_{SC}$ = 1V, $V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		-	-	1.1	V
Short-Circuit Trip Level	V <sub>SC(ref)</sub>	V <sub>CC</sub> = 15V (Note 5)			0.51	0.56	V
Sensing Voltage of IGBT Current	V <sub>SEN</sub>	$R_{SC}$ = 56 $\Omega$ , $R_{SU}$ = $R_{SV}$ = $R_{SW}$ = 0 $\Omega$ and $I_{C}$ = 15A (Note Fig. 7)			0.51	0.56	V
Supply Circuit Under-	UV <sub>CCD</sub>	Detection Level		11.5	12	12.5	V
Voltage Protection	UV <sub>CCR</sub>	Reset Level		12	12.5	13	V
	UV <sub>BSD</sub>	Detection Level		7.3	9.0	10.8	V
	UV <sub>BSR</sub>	Reset Level		8.6	10.3	12	V
Fault Output Pulse Width	t <sub>FOD</sub>	C <sub>FOD</sub> = 33nF (Note 6	)	1.4	1.8	2.0	ms
ON Threshold Voltage	V <sub>IN(ON)</sub>	High-Side	Applied between IN <sub>(UH)</sub> , IN <sub>(VH)</sub> ,	-	-	0.8	V
OFF Threshold Voltage	V <sub>IN(OFF)</sub>		IN <sub>(WH)</sub> - COM <sub>(H)</sub>	3.0	-	-	V
ON Threshold Voltage	V <sub>IN(ON)</sub>	Low-Side	Applied between IN <sub>(UL)</sub> , IN <sub>(VL)</sub> ,	-	-	0.8	V
OFF Threshold Voltage	V <sub>IN(OFF)</sub>		IN <sub>(WL)</sub> - COM <sub>(L)</sub>	3.0	-	-	V
Resistance of Thermistor	R <sub>TH</sub>	@ T <sub>TH</sub> = 25°C (Note Fig. 6) (Note 7)		-	50	-	kΩ
		@ T <sub>TH</sub> = 100°C (Note	e Fig. 6) (Note 7)	-	3.4	-	kΩ

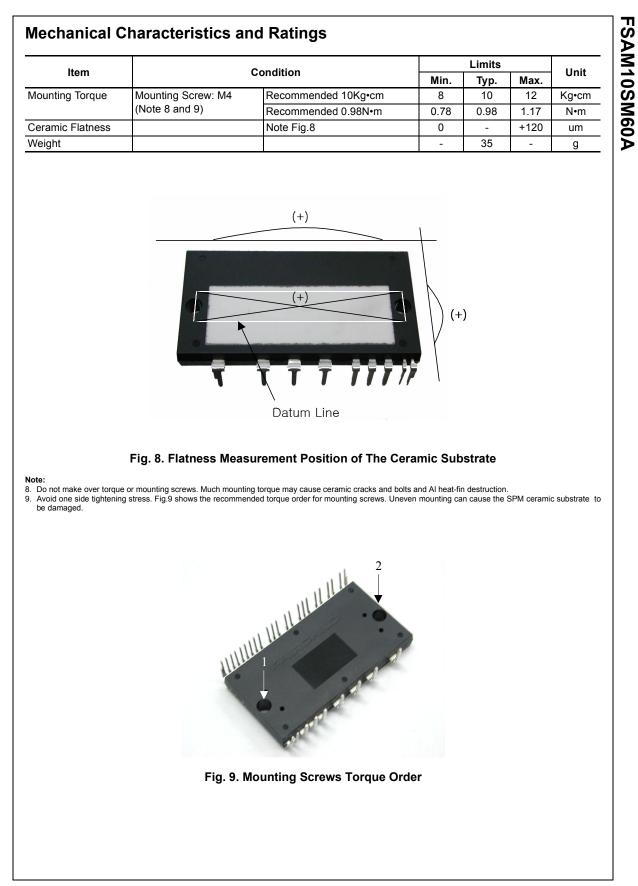
**Electrical Characteristics** (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

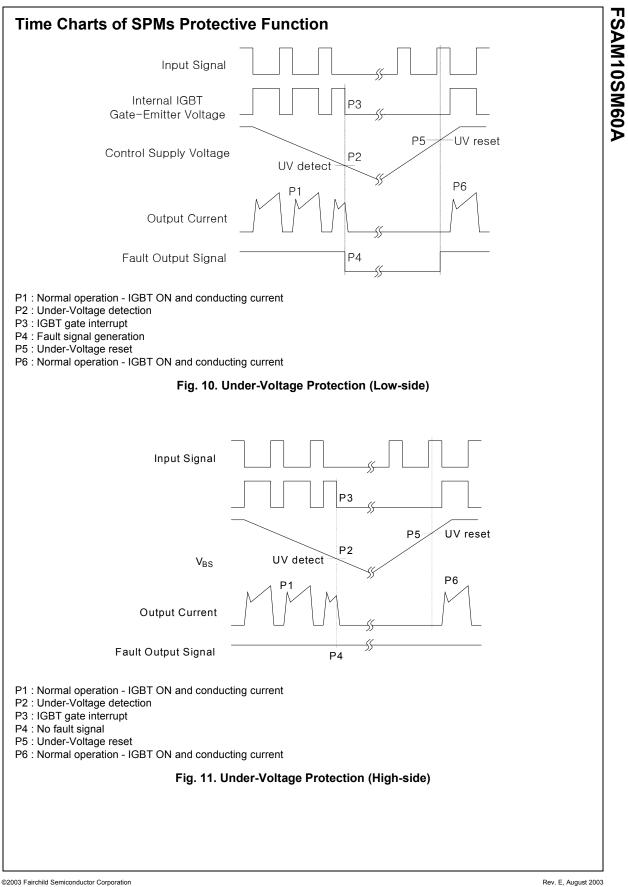
**Note:** 5. Short-circuit current protection is functioning only at the low-sides. It would be recommended that the value of the external sensing resistor ( $R_{SC}$ ) should be selected around 56  $\Omega$  in order to make the SC trip-level of about 15A at the shunt resistors ( $R_{SU}$ ,  $R_{SV}$ ,  $R_{SW}$ ) of  $0\Omega$ . For the detailed information about the relationship between the external sensing resistor ( $R_{SC}$ ) and the shunt resistors ( $R_{SU}$ ,  $R_{SV}$ ,  $R_{SW}$ ), please see Fig. 7. 6. The fault-out pulse width  $t_{FOD}$  depends on the capacitance value of  $C_{FOD}$  according to the following approximate equation :  $C_{FOD} = 18.3 \times 10^{-6} \times t_{FOD}[F]$ 7.  $T_{TH}$  is the temperature of thermistor itself. To know case temperature ( $T_{C}$ ), please make the experiment considering your application.

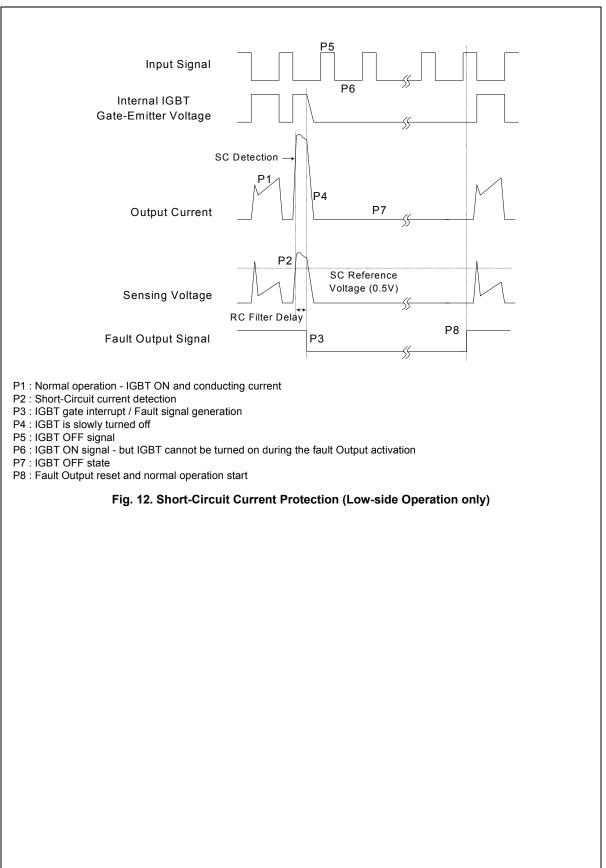
# **Recommended Operating Conditions**

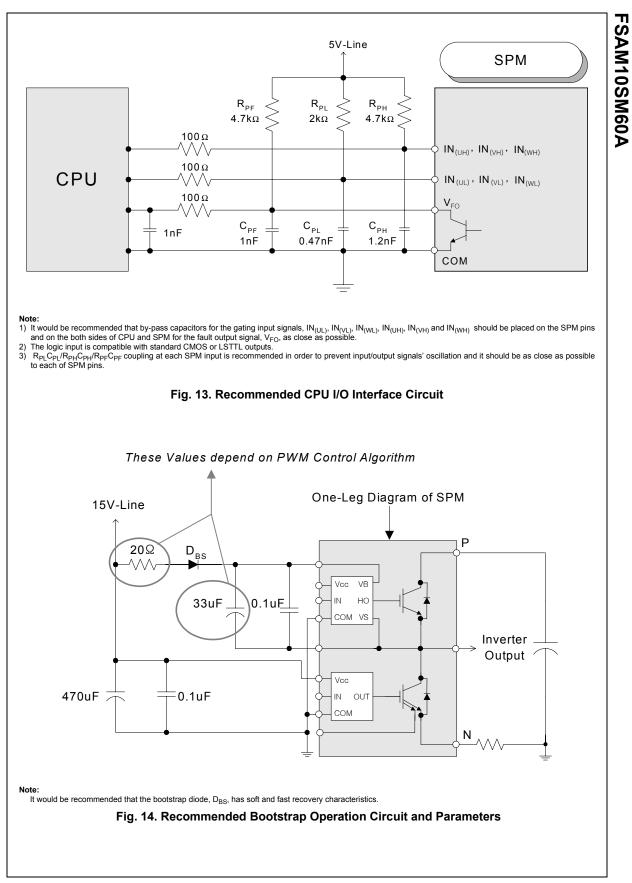
140.00	Symbol	Condition		Values		Unit	
ltem	Symbol	Condition	Min. Typ. Max		Max.		
Supply Voltage	V <sub>PN</sub>	Applied between P - N <sub>U</sub> , N <sub>V</sub> , N <sub>W</sub>		300	400	V	
Control Supply Voltage	V <sub>CC</sub>	$ \begin{array}{llllllllllllllllllllllllllllllllllll$		15	16.5	V	
High-side Bias Voltage	V <sub>BS</sub>	Applied between $V_{B(U)}$ - $V_{S(U)}$ , $V_{B(V)}$ - $V_{S(V)}$ , $Y_{B(W)}$ - $V_{S(W)}$		15	16.5	V	
Blanking Time for Preventing Arm-short	t <sub>dead</sub>	For Each Input Signal		-	-	us	
PWM Input Signal	f <sub>PWM</sub>	$T_C \le 100^{\circ}C, T_J \le 125^{\circ}C$	-	5	-	kHz	
Input ON Threshold Voltage	V <sub>IN(ON)</sub>	$      _{N(ON)}  \begin{array}{l} \mbox{Applied between IN}_{(UH)}, \mbox{ IN}_{(VH)}, \mbox{ IN}_{(WH)} - \\ \mbox{ COM}_{(H)}, \mbox{ IN}_{(UL)}, \mbox{ IN}_{(VL)}, \mbox{ IN}_{(WL)} - \mbox{ COM}_{(L)} \end{array} $		0~0.65	5	V	
Input OFF Threshold Voltage	V <sub>IN(OFF)</sub>	$\begin{array}{l} \text{Applied between IN}_{(\text{UH})}, \text{IN}_{(\text{VH})}, \text{IN}_{(\text{WH})} - \\ \text{COM}_{(\text{H})}, \text{IN}_{(\text{UL})}, \text{IN}_{(\text{VL})}, \text{IN}_{(\text{WL})} - \text{COM}_{(\text{L})} \end{array}$		4 ~ 5.5		V	

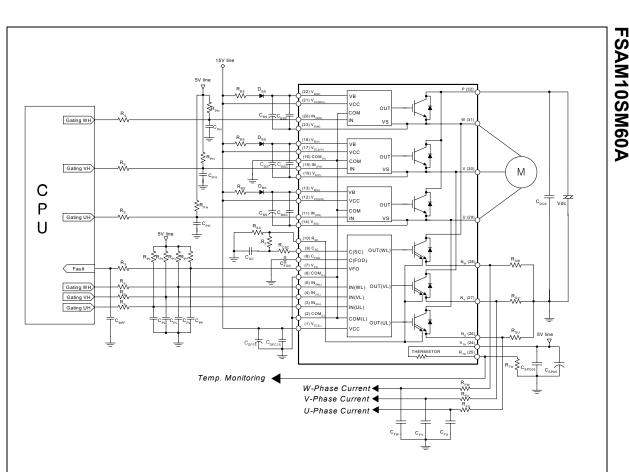








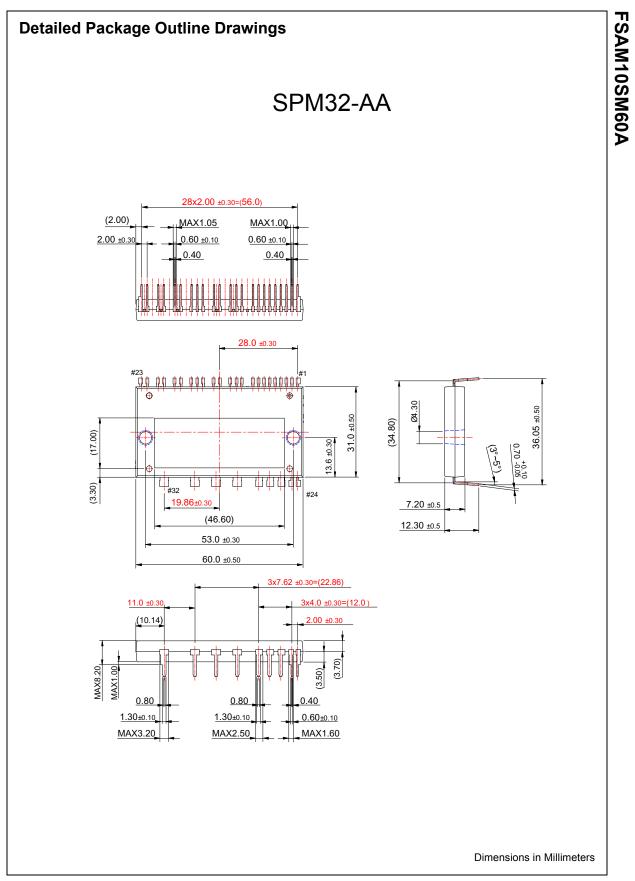




### Note:

- 1) RPLCPL/RPHCPH/RPFCPF coupling at each SPM input is recommended in order to prevent input signals' oscillation and it should be as close as possible to each SPM input pin.
- 2) By virtue of integrating an application specific type HVIC inside the SPM, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible
- 3) V<sub>FO</sub> output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7kΩ resistance. Please refer to Fig. 15. C<sub>SP15</sub> of around 7 times larger than bootstrap capacitor C<sub>BS</sub> is recommended.
- 5) V<sub>FO</sub> output pulse width should be determined by connecting an external capacitor(C<sub>FOD</sub>) between C<sub>FOD</sub>(pin8) and COM<sub>(L)</sub>(pin2). (Example : if C<sub>FOD</sub> = 33 nF, then t<sub>FO</sub> = 1.8 ms (typ.)) Please refer to the note 6 for calculation method.
- 6) Each input signal line should be pulled up to the 5V power supply with approximately  $4.7k\Omega$  (at high side input) or  $2k\Omega$  (at low side input) resistance (other RC coupling circuits at each input may be needed depending on the PWM control scheme used and on the wiring impedance of the system's printed circuit board). Approximately a 0.22~2nF by-pass capacitor should be used across each power supply connection terminals.
- To prevent errors of the protection function, the wiring around  $R_{SC}$ ,  $R_{F}$  and  $C_{SC}$  should be as short as possible. 8) In the short-circuit protection circuit, please select the  $R_{F}C_{SC}$  time constant in the range  $3 \sim 4 \, \mu$ . 9) To enhance the noise immunity,  $C_{SC}$  pin should be connected to the external circuit through a series resistor,  $R_{CSC}$ , which is approximately 390 $\Omega$ .  $R_{SCS}$  should be connected to  $C_{\mbox{\scriptsize SC}}$  pin as close as possible.
- 10)Each capacitor should be mounted as close to the pins of the SPM as possible.
- 11)To prevent surge destruction, the wiring between the smoothing capacitor and the P&N pins should be as short as possible. The use of a high frequency noninductive capacitor of around 0.1~0.22 uF between the P&N pins is recommended.
- 12)Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays. It is recommended that the distance be 5cm at least.

### Fig. 15. Typical Application Circuit



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EcoSPARK™	HiSeC™	MSXPro™	RapidConfigure™	UHC™
E <sup>2</sup> CMOS <sup>™</sup>	I <sup>2</sup> C <sup>™</sup>	OCX™	RapidConnect™	UltraFET <sup>®</sup>
EnSigna™	ImpliedDisconnect™	OCXPro™	SILENT SWITCHER®	VCX™
FACT™	ISOPLANAR™	<b>OPTOLOGIC<sup>®</sup></b>	SMART START™	
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The Power Fran		PACMAN™	Stealth™	
Programmable A		POP™	SuperSOT™-3	

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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

# **PRODUCT STATUS DEFINITIONS**

**Definition of Terms** 

Product Status	Definition
Formative or In Design	This datasheet contains the design specifications for product development. Specifications may change in any manner without notice.
First Production	This datasheet contains preliminary data, and supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice in order to improve design.
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